

# Exam Winter Semester 2022

## Student Group

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### Exercise E1 Resistance of a Wire by Resistivity (written test, approx. 6 % of a 60-minute written test, WS2022)

2. Heating elements made of solid nichrome wire with a diameter of  $d = 0.357 \text{ mm}$  and a length of  $l = 3 \text{ m}$  are used for heating elements. The electric power dissipation (= heat flow) of  $P = 40 \text{ W}$  is necessary.

Determine the current  $I$  and the operating voltage  $U$  for heating elements.

The Nichrome wire has a resistivity of  $\rho = 1.10 \cdot 10^{-6} \text{ } \Omega \cdot \text{m}$ .

The heating element is  $l = 3 \text{ m}$  long and has a diameter of  $d = 0.357 \text{ mm}$ .

∴ Calculate the resistance  $R$  of the heating element.

Solution

$$\begin{aligned} P &= U \cdot I = R \cdot I^2 \quad \rightarrow \quad I = \\ &= \sqrt{\frac{P}{R}} = \sqrt{\frac{40 \text{ W}}{0.33 \text{ } \Omega}} \end{aligned}$$

$$\begin{aligned} R &= \rho \cdot \frac{l}{A} \quad \text{with } A = r^2 \cdot \pi = \\ &= \frac{1}{4} d^2 \cdot \pi \quad \parallel \quad R = \rho \cdot \frac{l}{\frac{1}{4} d^2 \cdot \pi} \quad \parallel \quad R = \\ &= 1.10 \cdot 10^{-6} \text{ } \Omega \cdot \text{m} \cdot \frac{4 \cdot 3 \text{ m}}{(3.57 \cdot 10^{-3} \text{ m})^2 \cdot \pi} \end{aligned}$$

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### Exercise E2 Temperature-dependent Resistance

**(written test, approx. 6 % of a 60-minute written test, WS2022)**

2. A refrigerator exhibits a temperature coefficient of resistance in the refrigeration system. The circuit has a resistance of  $10 \text{ k}\Omega$  at  $25^\circ\text{C}$ . Your answer.

Its temperature coefficients are:  $\alpha = 0.01 \text{ K}^{-1}$  and  $\beta = 71 \cdot 10^{-6} \text{ K}^{-2}$

The temperature inside the refrigeration system can reach down to  $-40^\circ\text{C}$ .

$$R = 6.5 \text{ k}\Omega$$

The power transfer resistor  $P$  is a part of the circuit and generates heat. Therefore, a solution is to increase the heat flow up the refrigeration system.

Therefore, with constant  $U$  and increasing  $R$  the power decreases. Ten times more resistance decreases the heat flow to one-tenth.

$$R = R_0 \cdot (1 + \alpha \cdot \Delta T + \beta \cdot \Delta T^2) \quad | \quad \Delta T = T_{\text{end}} - T_{\text{start}} \\ R = 10 \text{ k}\Omega \cdot \left( (1 + 0.01 \text{ K}^{-1} \cdot (-40^\circ\text{C} - 25^\circ\text{C})) + 71 \cdot 10^{-6} \text{ K}^{-2} \cdot (-40^\circ\text{C} - 25^\circ\text{C})^2 \right)$$

**Exercise E2 Temperature-dependent Resistance**

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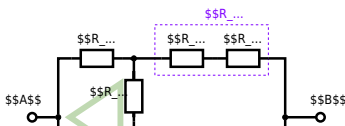
**Exercise E3 Pure Resistor Network Simplification**  
**(written test, approx. 13 % of a 60-minute written test, WS2022)**

The following shall be solved at once,  $R_2 = R_3 = 100 \Omega$  and the switch shall be closed. Calculate the equivalent resistance  $R_{eq}$  between  $A$  and  $B$ .

Solution

$$R_{eq} = 133.8 \Omega$$

Now a wye-delta transformation is necessary.



Since  $R_2 = R_3$  and based on the equations for the transformation, the transformed  $R_Y$  is given as:

$$R_Y = \frac{R_2 \cdot R_2}{R_2 + R_2 + R_2} = \frac{(100 \Omega)^2}{3 \cdot 100 \Omega} = \frac{1}{3} \cdot 100 \Omega = 33.33 \Omega$$

The equivalent resistor is given by a parallel configuration of resistors in series:

$$R_{eq} = R_Y + (R_Y + R_1 + R_1) \parallel (R_Y + R_2) \parallel (R_Y + R_2)$$

The switch shall now be open. Calculate the equivalent resistance  $R_{eq}$  between  $A$  and  $B$ .

Solution



The equivalent resistor is given by a parallel configuration of resistors in series:

$$R_{\text{eq}} = (R_2 + R_1 + R_1) \parallel (R_2 + R_2) \parallel (100 \Omega + 200 \Omega + 200 \Omega) \parallel (100 \Omega + 100 \Omega) \parallel (500 \Omega) \parallel (200 \Omega) \parallel (500 \Omega \cdot 200 \Omega) / (500 \Omega + 200 \Omega)$$

**Exercise E4 Pure Resistor Network Simplification (written test, approx. 13 % of a 60-minute written test, WS2022)**

The following shall be solved with  $R_1 = 200 \Omega$ ,  $R_2 = R_3 = 100 \Omega$  and the source  $B = 15 \text{ V}$ .  
 Result given:  $R_{\text{eq}} = B$ .

Solution

$$R_{\text{eq}} = 132.8 \Omega$$

Now a wye-delta transformation is necessary.



Since  $R_2=R_3$  and based on the equations for the transformation, the transformed  $R_Y$  is given as: 
$$R_Y = \frac{R_2 \cdot R_2}{R_2 + R_2 + R_2} = \frac{(100 \Omega)^2}{3 \cdot 100 \Omega} = \frac{1}{3} \cdot 100 \Omega = 33.33 \Omega$$

The equivalent resistor is given by a parallel configuration of resistors in series: 
$$R_{eq} = R_Y + (R_Y + R_1 + R_1) \parallel (R_Y + R_2) \parallel R_{eq} = 33.33 \Omega + (33.33 \Omega + 400 \Omega) \parallel (33.33 \Omega + 100 \Omega)$$

1. The switch shall now be open. Calculate the equivalent resistance  $R_{eq}$  between A and B.

Solution



The equivalent resistor is given by a parallel configuration of resistors in series:

$$R_{\text{eq}} = (R_2 + R_1 + R_{-1}) \parallel (R_2 + R_2) \parallel R_{\text{eq}} = (100 \Omega + 200 \Omega + 200 \Omega) \parallel (100 \Omega + 100 \Omega) \parallel R_{\text{eq}} = (500 \Omega) \parallel (200 \Omega) \parallel R_{\text{eq}} = \frac{500 \Omega \cdot 200 \Omega}{500 \Omega + 200 \Omega} \parallel$$

**Exercise E5 Equivalent linear Source  
(written test, approx. 14 % of a 60-minute written test, WS2022)**

The circuit in the following has to be simplified.  
Result

$$U_{\text{s}} = U_{\text{AB}} = 4.5 \text{ V} \quad R_{\text{i}} = R_{\text{AB}} = 6 \Omega$$



Calculate the internal resistance  $R_{in}$  and the source voltage  $U_{s}$  of an equivalent linear voltage source on the connectors  $A$  and  $B$ .  $R_1=5.0 \Omega$ ,  $U_2=6.0 \text{ V}$ ,  $R_3=10 \Omega$ ,  $I_4=4.2 \text{ A}$ ,  $R_5=10 \Omega$ ,  $R_6=7.5 \Omega$ ,  $R_7=15 \Omega$ . Use equivalent sources in order to simplify the circuit!

### Solution

The best thing is to re-think the wiring like rubber bands and adjust them:



The linear voltage source of  $U_2$  and  $R_1$  can be transformed into a current source  $I_2 = \frac{U_2}{R_1}$  and  $R_1$ :



Now a lot of them can be combined. The resistors  $R_1$ ,  $R_3$ ,  $R_5$  are in parallel, like also  $I_2$  and  $I_4$ :

$$R_{135} = R_1 || R_3 || R_5$$

$$I_{24} = I_2 - I_4$$

The resulting circuit can again be transformed:



Here, the  $U_{24}$  is calculated by  $I_{24}$  as the following:

$$U_{24} = I_{24} \cdot R_{135} + I_2 \cdot R_2 + I_4 \cdot R_3 + I_5 \cdot R_4 + I_6 \cdot R_5$$

$$U_{24} = U_2 \cdot \frac{R_7}{R_6 + R_7 + R_1 || R_3 || R_5} - I_4 \cdot R_1 || R_3 || R_5$$

On the right side of the last circuit, there is a voltage divider given by  $R_{135}$ ,  $R_6$ , and  $R_7$ .

Therefore the voltage between  $A$  and  $B$  is given as:

$$U_{AB} = U_{24} \cdot \frac{R_7}{R_6 + R_7 + R_1 || R_3 || R_5} - I_4 \cdot \frac{R_7 \cdot R_1 || R_3 || R_5}{R_6 + R_7 + R_1 || R_3 || R_5}$$

For the internal resistance  $R_i$  the ideal voltage source is substituted by its resistance ( $=0\Omega$ , so a short-circuit):

$$R_{AB} = R_7 || (R_6 + R_1 || R_3 || R_5)$$

with  $R_1 || R_3 || R_5 = 5\Omega || 10\Omega || 10\Omega = 5\Omega || 5\Omega = 2.5\Omega$ :

$$U_{AB} = \frac{6.0\text{V}}{5.0\Omega} - 4.2\Omega \cdot \frac{15\Omega \cdot 2.5\Omega}{7.5\Omega + 15\Omega + 2.5\Omega}$$

$$R_{AB} = 15\Omega || (7.5\Omega + 2.5\Omega)$$

### Exercise E6 Equivalent linear Source (written test, approx. 14 % of a 60-minute written test, WS2022)

The circuit in the following has to be simplified.  
Result

$$U_s = U_{AB} = 4.5\text{V} \quad R_i = R_{AB} = 6\Omega$$



Calculated the internal resistance  $R_{\text{int}}$  and the source voltage  $U_{\text{oc}}$  of an equivalent linear voltage source on the connectors  $\text{A}$  and  $\text{B}$ . 
$$R_1=5.0 \text{ } \Omega, \quad U_2=6.0 \text{ V}, \quad R_3= 10 \text{ } \Omega, \quad I_4=4.2 \text{ A}, \quad R_5=10 \text{ } \Omega, \quad R_6=7.5 \text{ } \Omega, \quad R_7=15 \text{ } \Omega$$
 Use equivalent sources in order to simplify the circuit!

Solution

The best thing is to re-think the wiring like rubber bands and adjust them:



The linear voltage source of  $U_2$  and  $R_1$  can be transformed into a current source  $I_2 = \frac{U_2}{R_1}$  and  $R_1$ :



Now a lot of them can be combined. The resistors  $R_1$ ,  $R_3$ ,  $R_5$  are in parallel, like also  $I_2$  and  $I_4$ :

$$R_{135} = R_1 || R_3 || R_5$$

$$I_{24} = I_2 - I_4$$

The resulting circuit can again be transformed:



Here, the  $U_{24}$  is calculated by  $I_{24}$  as the following:

$$U_{24} = I_{24} \cdot R_{135} + I_{24} \cdot R_5$$

$$U_{AB} = U_{24} \cdot \frac{R_7}{R_6 + R_7 + R_1 || R_3 || R_5} - I_4 \cdot R_1 || R_3 || R_5$$

On the right side of the last circuit, there is a voltage divider given by  $R_{135}$ ,  $R_6$ , and  $R_7$ .

Therefore the voltage between  $A$  and  $B$  is given as:

$$U_{AB} = U_{24} \cdot \frac{R_7}{R_6 + R_7 + R_1 || R_3 || R_5} - (U_2 \cdot \frac{R_1}{R_1 + R_3 + R_5} - I_4) \cdot R_1 || R_3 || R_5$$

For the internal resistance  $R_i$  the ideal voltage source is substituted by its resistance ( $=0 \Omega$ , so a short-circuit):

$$R_{AB} = R_7 || (R_6 + R_1 || R_3 || R_5)$$

with  $R_1 || R_3 || R_5 = 5 \Omega || 10 \Omega || 10 \Omega = 5 \Omega || 5 \Omega = 2.5 \Omega$ :

$$U_{AB} = \frac{6.0 \text{ V}}{5.0 \Omega} - 4.2 \text{ A} \cdot 2.5 \Omega$$

$$R_{AB} = 15 \Omega || (7.5 \Omega + 2.5 \Omega)$$

**Exercise E7 Charging Capacitors**  
**(written test, approx. 16 % of a 60-minute written test, WS2022)**

The circuit below is a RC circuit consisting of a DC voltage source  $U$ , a resistor  $R_1$ , a resistor  $R_2$ , a capacitor  $C$ , and a switch  $S_1$ . The switch  $S_1$  is initially open. The voltage across the capacitor is again  $0 \text{ V}$  at the moment  $t_0 = 0 \text{ s}$  when the switch  $S_1$  is closed. Calculate the voltage  $u_c(t_2)$  across the capacitor at  $t_2 = 1 \text{ ms}$  after closing the switch.

**Solution:** To solve this, first create an equivalent linear voltage source from  $U$ ,  $R_1$ , and  $R_2$ .

**Solution:** The ideal voltage source  $U_{eq}$  is given by  $U_{eq} = U \cdot \frac{R_2}{R_1 + R_2}$  and the internal resistance  $R_{eq}$  is given by  $R_{eq} = R_1 || R_2$ .

On an alternative view, one can try to create an equivalent linear voltage source again. Then, the internal resistance is given by substituting the ideal voltage source is again short-circuiting  $R_2$ .

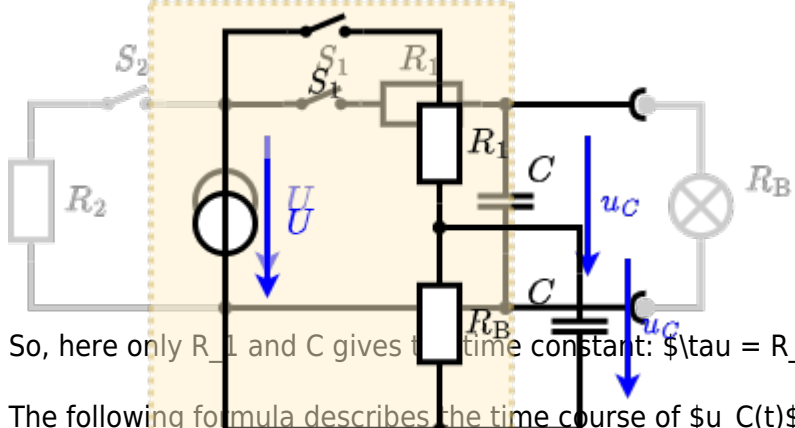


The circuit contains a voltage source  $U=12\text{ V}$ , a switch  $S_1$ , a resistor of  $R_1=20\text{ }\Omega$  and a capacitor of  $C=100\text{ }\mu\text{F}$ . The switch  $S_2$  to an additional consumer  $R_2$  will be considered to be open for the first task. At the moment  $t_0=0\text{ s}$  the switch  $S_1$  is closed, the voltage across the capacitor is  $u_c(t_0)=0\text{ V}$ .



First do not consider the light bulb - it is not connected to the RC circuit. Calculate the point of time  $t_1$  when  $u_c(t_1)=0.5 \cdot U$ .

Solution



So, here only  $R_1$  and  $C$  gives the time constant:  $\tau = R_1 \cdot C$

The following formula describes the time course of  $u_c(t)$  which has to be  $u_c(t_1)=0.5 \cdot U$ : 
$$u_c(t) = U \cdot (1 - e^{-t/\tau}) = 0.5 \cdot U$$
 It has to be rearranged to  $(1 - e^{-t/\tau}) = 0.5 \Rightarrow e^{-t/\tau} = 0.5 \Rightarrow t/\tau = \ln(0.5) \Rightarrow t = \tau \cdot \ln(0.5) = R_1 \cdot C \cdot \ln(0.5)$

Therefore, the voltage of the equivalent linear voltage source is:  $U_s = U \cdot \frac{R_B}{R_1 + R_B} = 1/2 \cdot U$  The internal resistance is given by substituting the ideal voltage source with its resistance ( $=0\text{ }\Omega$ , short-circuit). 
$$R_i = R_1 \parallel R_B = 10\text{ }\Omega$$

$$u_c(t_2) = U_s \cdot (1 - e^{-t_2/(R_i \cdot C)}) = \frac{1}{2} \cdot U \cdot (1 - e^{-1\text{ ms}/(10\text{ }\Omega \cdot 100\text{ }\mu\text{F})})$$

### Exercise E8 Charging Capacitors (written test, approx. 16 % of a 60-minute written test, WS2022)

The circuit (as shown in the figure) consists of a DC voltage source  $U=6\text{ V}$ , a resistor  $R_1=20\text{ }\Omega$ , a capacitor  $C=20\text{ }\mu\text{F}$  and a light bulb  $R_B=10\text{ }\Omega$ . The switch  $S_1$  is open. At the moment  $t_0=0\text{ s}$  when the switch  $S_1$  is closed. Calculate the voltage  $u_c(t_2)$  across the capacitor at  $t_2=1\text{ ms}$  after closing the switch.

**Solution** To solve this, first create an equivalent linear voltage source from  $U$ ,  $R_1$ , and  $R_B$ . 
$$U_s = U \cdot \frac{R_B}{R_1 + R_B} = 2\text{ V}$$
 
$$R_i = R_1 \parallel R_B = 13.3\text{ }\Omega$$

Solution

The ideal voltage source is  $U = 12 \text{ V}$ . The internal resistance is given by substituting the ideal voltage source is again short-circuiting  $R_2$ .

On an alternative view, one can try to create an equivalent linear voltage source again. Then, the internal resistance is given by substituting the ideal voltage source is again short-circuiting  $R_2$ .



The circuit contains a voltage source  $U = 12 \text{ V}$ , a switch  $S_1$ , a resistor of  $R_1 = 20 \text{ }\Omega$  and a capacitor of  $C = 100 \text{ }\mu\text{F}$ .

The switch  $S_2$  to an additional consumer  $R_2$  will be considered to be open for the first asks. At the moment  $t_0 = 0$  the switch  $S_1$  is closed, the voltage across the capacitor is  $u_c(t_0) = 0$ .

First do not consider the light bulb - it is not connected to the RC circuit. Calculate the point of time  $t_1$  when  $u_c(t_1) = 0.5 \cdot U$ .



Solution

An equivalent linear voltage source can be given with  $U$ ,  $R_1$ , and  $R_B$  as seen in yellow.

Therefore, the voltage of the equivalent linear voltage source is:  $U_s = U \cdot \frac{R_B}{R_1 + R_B} = 1/2 \cdot U$ . The internal resistance is given by substituting the ideal voltage source with its resistance ( $R = 0 \text{ }\Omega$ , short-circuit).  $R_i = R_1 \parallel R_B = 10 \text{ }\Omega$

$u_c(t_2) = U_s \cdot (1 - e^{-t_2/(R_i \cdot C)}) = \frac{1}{2} \cdot U \cdot (1 - e^{-t_2/(10 \text{ }\Omega \cdot 100 \text{ }\mu\text{F})})$   
So, here only  $R_1$  and  $C$  gives the time constant:  $\tau = R_1 \cdot C$

The following formula describes the time course of  $u_C(t)$  which has to be  $u_c(t_1) = 0.5 \cdot U$ :  $u_c(t) = U \cdot (1 - e^{-t/\tau}) = 0.5 \cdot U$   
It has to be rearranged to  $(1 - e^{-t/\tau}) = 0.5 \implies e^{-t/\tau} = 0.5 \implies -t/\tau = \ln(0.5) \implies t = \tau \cdot \ln(0.5)$

### Exercise E9 Analyzing complex Impedances (written test, approx. 14 % of a 60-minute written test, WS2022)

2. Given that the phasor voltage  $\underline{u}(t) = 50 \cos(300t - 10^\circ) \text{ V}$  and the phasor current  $\underline{i}(t) = 0.24 \cos(300t + \varphi) \text{ A}$  are both through the components ( $R$  and  $X_L$ ) shall be given.

After analysis, the full width dimensioned complex impedance  $\underline{Z}$  shall be extracted and the magnitude  $|Z|$  and phase  $\varphi$  shall be given.  $\underline{Z} = (2 + j4) \Omega$

.. Calculate the physical values of the two components.  $R = 2 \Omega, L = 4 \text{ mH}$

Solution

$$\underline{I} = \frac{\underline{U}}{\underline{Z}} \quad \text{with } \underline{U} = 50 \angle -10^\circ \text{ V, } \underline{Z} = (2 + j4) \Omega$$

The current  $I$  is  $0.24 \text{ A}$  and the phase angle is  $\varphi = 10^\circ - 90^\circ = -80^\circ$  (real)

resulting in  $\underline{Z} = \frac{\underline{U}}{\underline{I}} = \frac{50 \angle -10^\circ}{0.24 \angle -80^\circ} = 208.33 \angle 70^\circ \Omega$

Therefore, the component  $R = 208.33 \cos(70^\circ) = 72.5 \Omega$  and  $X_L = 208.33 \sin(70^\circ) = 194.1 \Omega$

Impedance  $\underline{Z} = R + jX_L = 72.5 + j194.1 \Omega$

With the complex part  $\varphi = \arctan\left(\frac{X_L}{R}\right) = \arctan\left(\frac{194.1}{72.5}\right) = 69^\circ$

The phase  $\varphi$  shall be calculated as  $\varphi = \arctan\left(\frac{\text{Im}(\underline{Z})}{\text{Re}(\underline{Z})}\right) = \arctan\left(\frac{194.1}{72.5}\right) = 69^\circ$

### Exercise E10 Analyzing complex Impedances (written test, approx. 14 % of a 60-minute written test, WS2022)

2. Given that the phasor voltage  $\underline{u}(t) = 50 \cos(300t - 10^\circ) \text{ V}$  and the phasor current  $\underline{i}(t) = 0.24 \cos(300t + \varphi) \text{ A}$  are both through the components ( $R$  and  $X_L$ ) shall be given.

After analysis, the full width dimensioned complex impedance  $\underline{Z}$  shall be extracted and the magnitude  $|Z|$  and phase  $\varphi$  shall be given.  $\underline{Z} = (2 + j4) \Omega$

.. Calculate the physical values of the two components.  $R = 2 \Omega, L = 4 \text{ mH}$

Solution

$$\underline{I} = \frac{\underline{U}}{\underline{Z}} \quad \text{with } \underline{U} = 50 \angle -10^\circ \text{ V, } \underline{Z} = (2 + j4) \Omega$$

The current  $I$  is  $0.24 \text{ A}$  and the phase angle is  $\varphi = 10^\circ - 90^\circ = -80^\circ$  (real)

resulting in  $\underline{Z} = \frac{\underline{U}}{\underline{I}} = \frac{50 \angle -10^\circ}{0.24 \angle -80^\circ} = 208.33 \angle 70^\circ \Omega$

Therefore, the component  $R = 208.33 \cos(70^\circ) = 72.5 \Omega$  and  $X_L = 208.33 \sin(70^\circ) = 194.1 \Omega$

Impedance  $\underline{Z} = R + jX_L = 72.5 + j194.1 \Omega$

With the complex part  $\varphi = \arctan\left(\frac{X_L}{R}\right) = \arctan\left(\frac{194.1}{72.5}\right) = 69^\circ$

The phase  $\varphi$  shall be calculated as  $\varphi = \arctan\left(\frac{\text{Im}(\underline{Z})}{\text{Re}(\underline{Z})}\right) = \arctan\left(\frac{194.1}{72.5}\right) = 69^\circ$

The absolute value of the impedance is given by 
$$|Z| = \sqrt{R^2 + \left(\frac{1}{\omega C} - \omega L\right)^2}$$
 with  $R = 1.00 \Omega$ ,  $C = 40 \text{ nF}$ ,  $L = 4.68 \mu\text{H}$ , and  $\omega = 2\pi \cdot 4 \text{ MHz}$ .  
 With the complex part comes the physical value: 
$$\varphi = \arctan\left(\frac{X_L - X_C}{R}\right) = \arctan\left(\frac{\omega L - \frac{1}{\omega C}}{R}\right)$$

**Exercise E11 Impedances at different Frequencies**  
 (written test, approx. 18 % of a 60-minute written test, WS2022)

At a series circuit with a resistor  $R_1 = 1.00 \Omega$ , a capacitor  $C_1 = 40 \text{ nF}$ , and an inductor  $L_1 = 4.68 \mu\text{H}$  in series, the current  $I = 10 \text{ mA}$  flows through the circuit. The resistor  $R_1$  shall have the same absolute value of the impedance as a capacitor  $C_2 = 40 \text{ nF}$  at  $f_2 = 4 \text{ MHz}$ .

**Solution**

The impedance of the resistor is  $R_1 = 1.00 \Omega$ .

The impedance of the capacitor is  $X_{C2} = \frac{1}{\omega_2 C_2} = \frac{1}{2\pi \cdot 4 \text{ MHz} \cdot 40 \text{ nF}} = -j0.995 \Omega$ .

The impedance of the inductor is  $X_{L1} = \omega_1 L_1 = 2\pi \cdot 4 \text{ MHz} \cdot 4.68 \mu\text{H} = j117.6 \Omega$ .

The total impedance of the series circuit is  $Z = R_1 + X_{L1} + X_{C2} = 1.00 \Omega + j117.6 \Omega - j0.995 \Omega = 1.00 \Omega + j116.6 \Omega$ .

The absolute value of the impedance is  $|Z| = \sqrt{1.00^2 + 116.6^2} = 116.6 \Omega$ .

The current is  $I = 10 \text{ mA}$ , so the voltage across the resistor is  $U_{R1} = I \cdot R_1 = 10 \text{ mA} \cdot 1.00 \Omega = 10 \text{ mV}$ .

The voltage across the capacitor is  $U_{C2} = I \cdot |X_{C2}| = 10 \text{ mA} \cdot 0.995 \Omega = 9.95 \text{ mV}$ .

The voltage across the inductor is  $U_{L1} = I \cdot |X_{L1}| = 10 \text{ mA} \cdot 117.6 \Omega = 1.176 \text{ V}$ .

**Exercise E12 Impedances at different Frequencies**  
 (written test, approx. 18 % of a 60-minute written test, WS2022)



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**Exercise E14 Complex Impedance Circuit**  
**(written test, approx. 15 % of a 60-minute written test, WS2022)**

1. Calculate the current  $i(t)$  through the resistor  $R$  in the circuit shown in Fig. 1. The voltage source is  $u(t) = 3.0 \cdot \sin(2\pi \cdot 15 \cdot t)$  V. The circuit consists of a resistor  $R = 10 \Omega$ , an inductor  $L = 330 \mu\text{H}$ , and a capacitor  $C = 0.22 \mu\text{F}$ , all in series.

```

Result
.. \begin{align*} Z = 19.8 - j48.2 \text{ } \Omega \quad |Z| = 48.2 \text{ } \Omega \quad \phi = 19.8 \text{ } \Omega \\ \end{align*}
\begin{align*} Z &= \frac{\hat{U}}{\hat{I}} \quad \hat{I} = \frac{\hat{U}}{Z} \\ Z_C &= \frac{1}{2\pi \cdot f \cdot C} = \frac{1}{2\pi \cdot 15 \text{ kHz} \cdot 0.22 \text{ } \mu\text{F}} \\ \end{align*}
With: \begin{align*} \hat{I} = \frac{\hat{U}}{|Z|} = \frac{3 \text{ V}}{48.2 \text{ } \Omega} = 0.062 \text{ A} \\ \end{align*}
\begin{align*} \hat{I} &= \frac{3 \text{ V}}{48.2 \text{ } \Omega} = 0.062 \text{ A} \\ \end{align*}
\underline{Z} = R + \underline{Z}_L + \underline{Z}_C = R + j \cdot \omega L - j \cdot \frac{1}{\omega C} \\ = 10 \text{ } \Omega + j \cdot 2\pi \cdot 15 \text{ kHz} \cdot 330 \text{ } \mu\text{H} - j \cdot \frac{1}{2\pi \cdot 15 \text{ kHz} \cdot 0.22 \text{ } \mu\text{F}} \\ = 10 \text{ } \Omega + j48.2 \text{ } \Omega - j19.8 \text{ } \Omega = 10 \text{ } \Omega + j28.4 \text{ } \Omega \\ |Z| = \sqrt{R^2 + (\underline{Z}_L - \underline{Z}_C)^2} = \sqrt{10^2 + 28.4^2} = 48.2 \text{ } \Omega \\ \phi = \arctan\left(\frac{\underline{Z}_L - \underline{Z}_C}{R}\right) = \arctan\left(\frac{28.4}{10}\right) = 19.8^\circ \\ \end{align*}

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Last update: **2023/04/01 23:19**

